A LATTICE PREDICTOR BASED ADAPTIVE VOLterra filter and a SYNCHRONIZED LEARNING ALGORITHM

Kenji Nakayama  Akihiro Hirano  Hiroaki Kashimoto

Dept. of Information and Systems Eng., Faculty of Eng., Kanazawa Univ. 2-40-20, Kodatsuno, Kanazawa, 920-8667, Japan  phone: +81 76 234 4896, fax: +81 234 4900, email: nakayama@t.kanazawa-u.ac.jp

ABSTRACT

This paper proposes a lattice predictor based adaptive Volterra filter and a synchronized learning algorithm. In the adaptive Volterra filter (AVF), the eigenvalue spread of a correlation matrix is extremely amplified, and its convergence is very slow for gradient methods. A lattice predictor is employed for whitening the input signal. Its convergence property is analyzed. The proposed learning algorithm is synchronously applied for updating the reflection coefficients. Furthermore, a synchronization learning algorithm is proposed, by which fast convergence and a small residual error can be achieved. Computer simulations, using colored signals and real speech signals, demonstrate that the proposed method is 10 times as fast as the DCT based AVF.

1. INTRODUCTION

Loud speakers in audio systems and small speakers embedded in a mobile phone have some nonlinearity. When they are used in a remote conference system and a visual phone, in which some echo are caused, nonlinear echo cancellers are very important.

Adaptive Volterra filters are one of hopeful candidates [1],[2],[3],[4]. It can express general nonlinearity. However, the Volterra polynomial has a huge number of terms, and the same number of filter coefficients are required. Furthermore, when the input signal is colored, the eigenvalue spread of a correlation matrix is extremely amplified, and convergence is very slow for gradient methods.

Many kinds of fast and stable learning algorithms for adaptive Volterra filters have been proposed [5],[3]. RLS algorithm is insensitive to the eigenvalue spread, at the expense of $O(N^2)$ computations. Another method is to combine a whitening process and an adaptive FIR Volterra filter. The Discrete Cosine Transform (DCT) has been applied to the whitening process [7]. Furthermore, an error surface and convergence property have been analyzed [8]. The DCT is not sufficient for the whitening process. A linear FIR predictor based on an AR model of the signal is good for whitening. However, it requires some time delay, and cannot be applied to some applications [6].

In this paper, in order to improve the whitening process without any time delay, a lattice predictor is employed for a whitening process. Furthermore, a synchronizing learning algorithm for updating the reflection coefficients and the filter coefficients is proposed. Its convergence property is analyzed. The proposed and conventional methods are compared with each other through computer simulations by using colored signals and real speech signals.

2. ADAPTIVE FIR VOLterra filter

2.1 Structure of AVF

Figure 1 shows a block diagram of an adaptive FIR Volterra filter (AVF). When a second-order Volterra polynomial is used, the output $y(n)$ is given by:

$$y(n) = \sum_{i=0}^{N-1} w_1(i)x(n-i) + \sum_{j=0}^{N-1} \sum_{k=0}^{N-1} w_2(j,k)x(n-j)x(n-k)$$

(1)

$$w_2(j,k) = w_2(k,j)$$

(2)

2.2 Eigenvalue Spread Amplification

The eigenvalue spread $\chi = \lambda_{max}/\lambda_{min}$ of the input signal $x(n)$ is extremely amplified by transforming it through the Volterra polynomial. Examples are shown in Table 1. The colored input signal $x(n)$ with $\chi = 780.9$ is transferred to the Volterra polynomial terms having $\chi = 657100$, which is 841 times as large as that of $x(n)$. For this reason, convergence of the adaptive Volterra filter is very slow for gradient methods.

2.3 Whitening Input Signal

2.3.1 Discrete Cosine Transform

Figure 2 shows the DCT with normalization [7]. The outputs of the tap delay line $x(n) = [x(n), x(n-1), ..., x(n-N+1)]$ are transformed through the DCT to...
\( q(n) = [q_0(n), q_1(n), \ldots, q_{N-1}(n)] \), and they are normalized by its standard deviation \( \sigma_{q,i} \). The outputs \( s_i(n) \) of this block are applied to the Volterra polynomial generating 1st-order and high-order terms. These terms are multiplied by filter coefficients, and are accumulated, resulting in the final output \( y(n) \). The DCT does not need any time delay.

Figure 2: DCT whitening process with normalization.

### 2.3.2 Linear FIR Predictor Based on AR Signal Model

When the signal can be modeled by the output of an AR circuit driven by the white noise, a linear FIR prediction error filter shown in Fig.3 is good for whitening. \( e(n) \) is used as the AVF input.

Figure 3: Linear predictor based on FIR filter.

### 2.4 Position of Whitening in Nonlinear Filters

Two kinds of positions for the whitening are shown in Figs.4 and 5, and are denoted Type-A and Type-B, respectively. Any time delay is not allowed in Type-A. Reason can be explained as follows: Suppose the linear predictor shown in Fig.3 is used in Type-A, and both the unknown system and the AVF have the \( N \)-th order FIR filter and a 2nd-order Volterra polynomial. The output of the unknown system includes \( x(n-i), i = 0, 1, \ldots, N-1 \) and their high-order terms. On the other hand, in the AVF part, \( e(n) \), which consists of \( x(n-i), i = 0, 1, \ldots, L-1 \), is the input of the AVF, then \( x(n-i), i = 0, 1, \ldots, N+L-1 \) and their high-order terms are included in its output signal. Therefore, their transfer functions are inherently different.

On the other hand, in Type B, any kinds of the whitening process, with or without time delay, can be employed. However, since the output of the unknown system may be sound from a loud speaker located in a conference room, it cannot be applied to some applications, such as echo cancellation.

Figure 4: Whitening adaptive filter input (Type-A).

Figure 5: Whitening both adaptive filter and unknown system inputs (Type-B).

### 3. LATTICE PREDICTOR BASED AVF

#### 3.1 Circuit Structure

In practical applications, Type A is important. Therefore, we employ the lattice predictor [9] for the whitening process. The proposed lattice predictor based AVF is shown in Fig.6. Under the conditions the delay line order is \( N \), the order of the Volterra polynomial is \( M \), the order of the lattice predictor is \( L \), and \( N > L \), the order of the transfer function \( Y(z)/X(z) \), and the number of filter coefficients in both AVF in Fig.1 and Fig.6 are the same. If the unknown system can be modeled by using the FIR Volterra filter, the same transfer function can be realized by the lattice predictor based AVF.

#### 3.2 Reflection Coefficient Update

The reflection coefficients are updated by the following equations [9].

\[
\kappa_m(n) = -\frac{2E[b_{m-1}(n-1)f_{m-1}(n)]}{E[|f_{m-1}(n)|^2 + |b_{m-1}(n-1)|^2]} \tag{3}
\]

\[
\kappa_{N,m}(n) = \gamma \kappa_{N,m}(n) + b_{m-1}(n-1)f_{m-1}(n) \tag{4}
\]

\[
\kappa_{D,m}(n) = \gamma \kappa_{D,m}(n) + |f_{m-1}(n)|^2 + |b_{m-1}(n-1)|^2 \tag{5}
\]

\[
0 < \gamma < 1
\]

\[
\kappa_m(n) = -\frac{\kappa_{N,m}(n)}{\kappa_{D,m}(n)} \tag{6}
\]
3.3 Synchronized Learning Algorithm

3.3.1 Lattice Predictor Based Adaptive Filters

Convergence property of the lattice predictor based FIR adaptive filter has been analyzed, and the synchronized learning algorithm has been proposed [10]. Updating the reflection coefficients and the filter coefficients are not synchronized, and some error remain.

The synchronizing method [10] is described here. The linear adaptive filter with the lattice predictor is equivalent to the circuit shown in Fig.6, except for the Volterra polynomial block. The filter coefficients $\mathbf{w}(n)$ is directly connected to $\mathbf{b}(n)$. The output $y(n)$ is

$$
\begin{align*}
\mathbf{b}(n) &= \mathbf{K}(n)\mathbf{x}(n) \\
y(n) &= \mathbf{w}^T(n)\mathbf{b}(n)
\end{align*}
$$

$b(n)$ is a vector of the backward prediction error $b_m(n)$, $\mathbf{K}(n)$ is a matrix consists of the reflection coefficients, $\mathbf{x}(n)$ is the input, $\mathbf{w}(n)$ is the filter coefficients. In the next iteration step, $\mathbf{K}(n)$ is updated to $\mathbf{K}(n+1)$, and $y(n+1)$ and $e(n+1)$ are generated by using $\mathbf{K}(n+1)$ and $\mathbf{w}(n)$. However, $\mathbf{w}(n)$ is optimized for $\mathbf{K}(n)$ not $\mathbf{K}(n+1)$. Therefore, $e(n+1)$ is not guaranteed to be reduced. For this reason, $\mathbf{w}(n)$ is modified as sa,

$$
\begin{align*}
\tilde{\mathbf{b}}(n+1) &= \mathbf{K}(n)\mathbf{x}(n+1) \\
\tilde{y}(n+1) &= \mathbf{w}^T(n)\tilde{\mathbf{b}}(n+1) \\
\mathbf{b}(n+1) &= \mathbf{K}(n+1)\mathbf{x}(n+1) \\
y(n+1) &= \mathbf{w}^T(n)\mathbf{b}(n+1)
\end{align*}
$$

$\tilde{y}(n+1)$ can reduce the output error. Therefore, the filter coefficients $\mathbf{w}(n)$ is modified as follows:

$$
\begin{align*}
\mathbf{K}^T(n+1)\tilde{\mathbf{w}}(n) &= \mathbf{K}^T(n)\mathbf{w}(n) \\
\tilde{\mathbf{w}}(n) &= \frac{\mathbf{K}^T(n)}{\mathbf{K}^T(n+1)}\mathbf{w}(n)
\end{align*}
$$

$\tilde{\mathbf{w}}(n)$ is used in the next iteration $n+1$, instead of $\mathbf{w}(n)$, for generating $\tilde{y}(n+1)$ and $\tilde{e}(n+1)$. The filter coefficients are updated to $\mathbf{w}(n+1)$ by using $\tilde{e}(n+1)$ and $\tilde{\mathbf{w}}(n)$. A combination of $\mathbf{K}(n)$ and $\mathbf{w}(n)$ is equivalent to that of $\mathbf{K}(n+1)$ and $\tilde{\mathbf{w}}(n)$. Thus, the output error is guaranteed to be decreased.

3.3.2 Lattice Predictor Based AVF

In the lattice predictor based AVF shown in Fig.6, the inputs of the Volterra function become $b_m(n), m < L \text{ and } b_{L-1}(n-m+L-1), L \leq m$. The filter coefficients $w_i(i)$ for the linear terms $b_i(n), i < L \text{ or } b_{L-1}(n+i-L-1), L \leq i$, can be modified by the same way as in Eq.(14). The other filter coefficients $w_2(j,k)$ for the 2nd-order terms, for instance $b_j(n)b_k(n)$, are modified in the following way.

$$
\begin{align*}
\mathbf{b}(n) &= \mathbf{K}(n)\mathbf{x}(n) \\
b_j(n) &= \mathbf{k}_j^T(n)\mathbf{x}(n) \\
b_j(n)b_k(n) &= (\mathbf{k}_j^T(n)\mathbf{x}(n)))(\mathbf{k}_k^T(n)\mathbf{x}(n))
\end{align*}
$$

$k_j(n)$ is the $j$-th row vector of $\mathbf{K}(n)$. The modification can be expressed as follows:

$$
\begin{align*}
\tilde{\mathbf{w}}_2(j,k)(n) &= \mathbf{k}_j^T(n)\mathbf{R}(n+1)\mathbf{k}_k(n) + 1 \\
\mathbf{w}_2(j,k)(n) &= \mathbf{k}_j^T(n)\mathbf{R}(n+1)\mathbf{k}_k(n) + 1
\end{align*}
$$

From this relation, the 2nd filter coefficients are modified as follows:

$$
\begin{align*}
\tilde{\mathbf{w}}_2(j,k)(n) &= \frac{\mathbf{k}_j^T(n)\mathbf{R}(n+1)\mathbf{k}_k(n)}{\mathbf{k}_j^T(n+1)\mathbf{R}(n+1)\mathbf{k}_k(n+1)}\mathbf{w}_2(j,k)(n)
\end{align*}
$$

4. SIMULATION AND DISCUSSIONS

4.1 Colored Signal

The colored signal is generated passing the white noise through a 2nd-order AR model, and is applied to the adaptive Volterra filter. The learning curves for the AVF without whitening, and with the DCT method in Type A and the linear predictor in Type B are shown in Fig.7. The NLMS algorithm and stepsize=0.1 are employed.
However, this type cannot be applied to echo canceler and so on.

Figure 8 shows the learning curves of the lattice predictor based AVF. Convergence property depends on a time constant $\gamma$, which control the reflection coefficient update. From Eqs. (4) and (5), when $\gamma < 1$ is very close to unity, the reflection coefficients $\kappa_i$ are very gradually adjusted. Until $\gamma = 0.999999$, the convergence can be improved. Compared with Fig. 7, the learning curves for the lattice predictor almost saturate around 50dB, however, it requires only 85,000 iterations until -40dB, while the DCT and the linear predictor need 320,000 iterations and 150,000 iterations, respectively.

4.2 Speech Signal

Speech signal is used as the input signal for the AVF. Figures 9 and 10 show the learning curves by using the DCT and the linear predictor, and by using the lattice predictor, respectively. In the former figure, the mean squared error in some interval is normalized by the mean squared signal in the same interval. In the latter case, it is normalized by the mean squared signal in the enter interval. So, the curves are different. From these results, the DCT cannot improve from the NLMS without whitening. The linear predictor can reach at -40dB with 250,000 iterations. On the other hand, the lattice predictor can reach at -40dB with 25,000 iterations. Thus, convergence time can be reduced to 1/10 of the FIR linear predictor.

5. CONCLUSIONS

In this paper, the lattice predictor based AVF and its synchronized learning algorithm have been proposed. Its convergence is dependent on the time constant parameter. Convergence can be improved to 10 times as fast as the conventional for stationary and nonstationary colored signals.

REFERENCES